

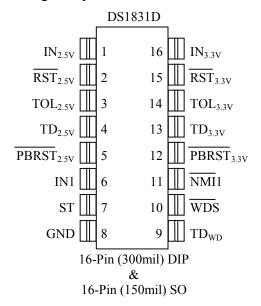
### DS1831C/D/E

# 3.3V/2.5V Multisupply MicroMonitor

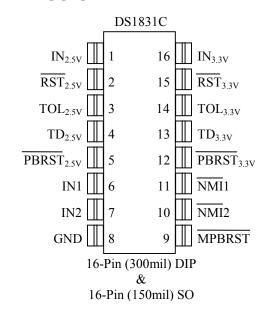
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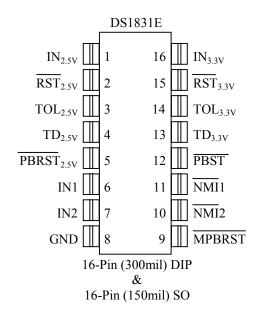
### **FEATURES**

- 2.5V power-on reset
- 3.3V power-on reset
- Two referenced comparators with separate outputs for monitoring additional supplies
- Internal power is drawn from higher of either the IN<sub>2.5V</sub> input or the IN<sub>3.3V</sub> input
- Excellent for systems designed to operate with multiple power supplies
- Asserts resets during power transients
- Pushbutton reset input for system override
- Maintains reset for user configurable times of 10ms, 100ms, or 1s
- Watchdog timer for software monitoring (DS1831D)
- Precision temperature-compensated voltage reference and voltage sensor
- 16-pin DIP and 16-pin 150mil SO available
- Operating Temperature of -40°C to +85°C



#### PIN ASSIGNMENT





### **DESCRIPTION**

The DS1831C multisupply monitor and reset monitors up to four system voltages: 2.5V supply, 3.3V (or 3V) supply, and two additional user configurable voltage monitors. DS1831 power for internal operation comes from the higher voltage level of the 3.3V input or the 2.5V input. One of these inputs must be greater than 1V for device operation. Pushbutton (manual reset) functionality is provided for the 2.5V

1 of 15 060303

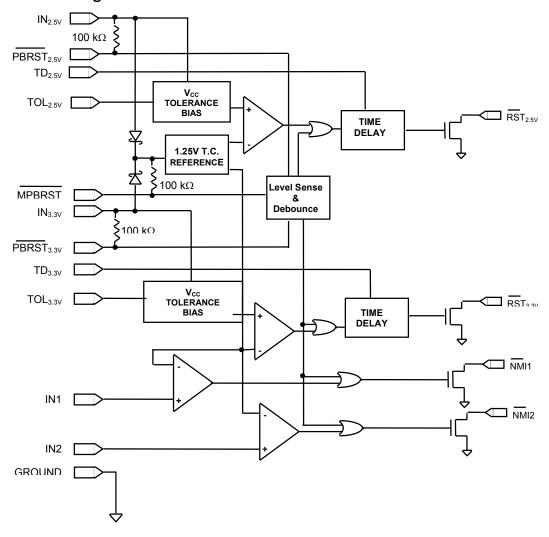
reset, the 3.3V reset or for all reset outputs by the master pushbutton. The DS1831D replaces one reference comparator and the master pushbutton with watchdog and the DS1831E replaces the 3.3V PBRST with a last reset status output.

TOL and TD inputs allow user configuration of the DS1831 for multiple applications. The TOL inputs configure the tolerance for the specified output and the TD inputs configure the reset time delays.

### **PIN DESCRIPTION**

IN <sub>2.5V</sub>	2.5V Power Supply Input	$\begin{array}{c} \text{MPBRST} \\ \text{TD}_{\text{WD}} \end{array}$	Master Pushbutton (DS1831C/E) Watchdog Time Delay Select (DS1831D)
RST <sub>2.5V</sub>	2.5V Reset Open Drain Output	$\frac{\overline{\text{NMI2}}}{\overline{\text{WDS}}}$	Non-maskable Interrupt 2 (DS1831C/E) Watchdog Status Output (DS1831D)
$TOL_{2.5V}$	Selects 2.5V Input Tolerance	NMI1	Non-maskable Interrupt 1
TD <sub>2.5V</sub>	Selects 2.5V Reset Time Delay	$\frac{\overline{PBRST}}{\overline{PBST}}_{3.3V}$	3.3V Reset Pushbutton (DS1831C/E) Pushbutton Status Output (DS1831E)
PBRST <sub>2.5V</sub>	2.5V Reset Pushbutton	$TD_{3.3V}$	Select 3.3V Reset Time Delay
IN1	Sense Input 1	$TOL_{3.3V}$	Selects 3.3V Input Tolerance
IN2 ST	Sense Input 2 (DS1831C/E) Watchdog Strobe Inputs (DS1831D)	RST <sub>3.3V</sub>	3.3V Reset Open Drain Output
GND	Ground	$IN_{3.3V}$	3.3V Power Supply Input

# **BLOCK DIAGRAM** Figure 1



#### OPERATION—POWER MONITOR

The DS1831 provides the functions of detecting out-of-tolerance conditions on a 3.3V (or 3V) and 2.5V power supply and warning a processor based system of impending power failure. When an input is detected as out-of-tolerance on either voltage input the  $\overline{\text{RST}}$  for that supply will be forced active low. When that input returns to a valid state the associated  $\overline{\text{RST}}$  will remain active for the time delay selected with the associated TD input and then return to an inactive state until the next input out-of-tolerance condition.

On power-up both resets are kept active for the selected reset time after the associated power supply input has reached the selected tolerance. This allows the power supply and system power to stabilize before  $\overline{RST}$  is released.

All internal operating current for the DS1831 will be supplied by either the  $IN_{3.3V}$  or  $IN_{2.5V}$  input which ever has the highest voltage level.

#### OPERATION—TOLERANCE SELECT

The DS1831 provides two TOL inputs for individual customization of the DS1831 to specific application requirements. If the TOL for the 2.5V supply is tied to the 2.5V input a 5% tolerance is selected. If the TOL is connected to ground a 10% tolerance is selected or if it is left unconnected a 15% tolerance is selected. If the TOL for the 3.3V supply is tied to the 3.3V input a 5% tolerance is selected, a 10% tolerance is selected if it is connected to ground, and a 20% tolerance is selected if the input is left unconnected. These tolerance conditions are set at power up and can only be changed by power cycling the device.

### **OPERATION—RESET TIME-DELAY SELECT**

The DS1831 provides two TD inputs for individual customization of reset time delays and an additional one for the DS1831D watchdog. TD inputs select time delays for the  $IN_{2.5V}$  and  $IN_{3.3V}$  resets outputs and the Watchdog on the DS1831D. The reset time delays are shown in Table 1. These allow the selection of minimum delays of 10ms, 100ms, and 1000ms.

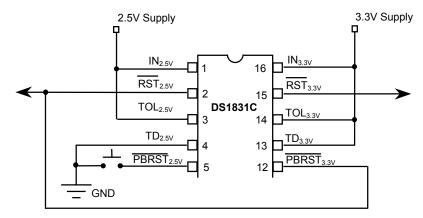
Wiring an individual reset output to the pushbutton input of the other voltage reset allows custom reset timings or allows for the sequencing of the reset outputs. See Figure 2.

These time delays are set at power-up and cannot be changed after the device reaches an in-tolerance condition.

TD INPLITS/RESET	AND WATCHDOG	TIME-DELAYS Table 1

	RESET TIME-DELAY			
TD	MIN	TYP	MAX	
GND	10ms	16ms	20ms	
Float	100ms	160ms	200ms	
$V_{CC}$	1000ms	1600ms	2000ms	

### **PUSHBUTTON RESET SEQUENCING Figure 2**



NOTE: The  $\overline{\text{RST}}_{2.5}$  volt output is connected to the  $\text{IN}_{3.3\text{V}}$  via a  $100\text{k}\Omega$  resistor in the pushbutton input and therefore does not require a pull-up resistor (an addition pull up can be used to accelerate responses.) If an external pull-up is used in this example it must be connected to the 3.3V power supply.

### **OPERATION—PUSHBUTTON RESET**

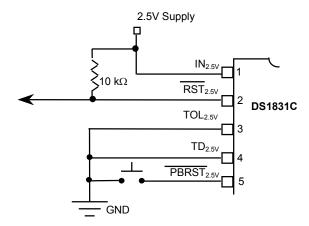
The DS1831 provides three pushbutton inputs for manual reset of the device. Pushbutton inputs for the 3.3V reset, 2.5V reset, and a master pushbutton reset (DS1831C and DS1831D only) input; provide multiple options for system control. The 3.3V pushbutton reset and 2.5V pushbutton resets provide a simple manual reset for the associated reset output; while the master pushbutton reset forces all resets and NMI outputs active low.

The 2.5V reset pushbutton input and the 3.3V reset pushbutton input provide manual reset control for each associated reset output. When the output associated with a pushbutton input is not active, a pushbutton reset can be generated by pulling the associated PBRST pin low for at least 20µs. When the pushbutton is held low the reset will be forced active and will remain active for a reset cycle after the pushbutton is released. See Figure 2 for an application example that allows a user to sequence the reset outputs.

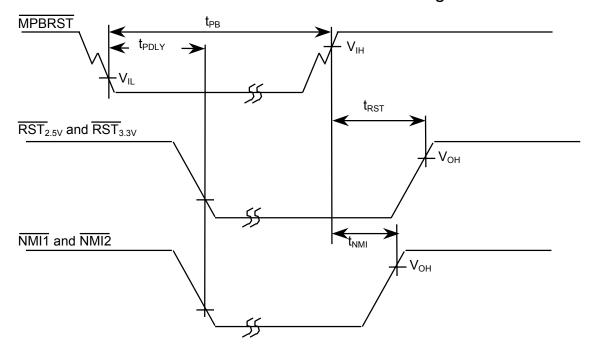
A master pushbutton reset cycle can be started if at least one voltage input (IN<sub>2.5V</sub>, IN<sub>3.3V</sub>, IN1, or IN2) is in tolerance and at least one output is active. A master pushbutton reset is generated by pulling the  $\overline{\text{MPBRST}}$  pin low for at least 20µs. When the pushbutton is held low all outputs are forced active and will remain active for a reset or NMI time delay after the pushbutton is released. The Master Pushbutton input is pulled high through an internal  $100\text{k}\Omega$  pull up resistor and debounced via internal circuitry. See Figure 3 for an application example. Figures 4 and 5 for the timing diagram.

The 2.5V and 3.3V pushbutton reset inputs are pulled high through an internal  $100k\Omega$  pull up resistor to the voltage input, which is associated with that pushbutton. The master pushbutton is pulled to the greater of the  $IN_{2.5V}$  and  $IN_{3.3V}$  inputs.

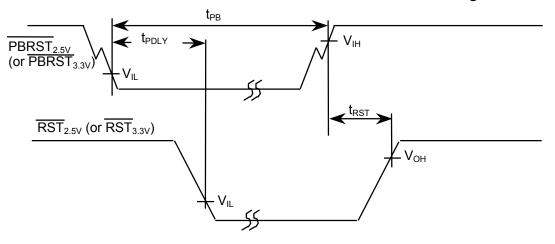
# **PUSHBUTTON RESET** Figure 3



### TIMING DIAGRAM—MASTER PUSHBUTTON RESET Figure 4



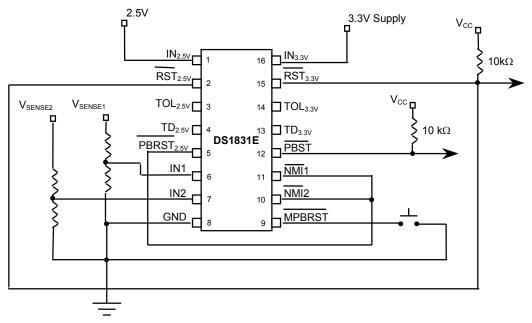
# TIMING DIAGRAM—2.5V OR 3.3V PUSHBUTTON RESET Figure 5



#### OPERATION—PUSHBUTTON STATUS

The DS1831E provides a master pushbutton status open drain output. The  $\overline{PBST}$  output indicates the status of the most recent reset condition. If the last reset was generated by the master pushbutton input it would maintain a low condition until cleared by another event (except the master pushbutton) generating a reset. Once cleared it will remain high until the master pushbutton is pulled low generating a reset condition. The  $\overline{PBST}$  output is open drain and will require a pull-up resistor on the output to maintain a valid condition. The value of the pull up resistor is not critical in most cases but must be set low enough to pull the output to a high state. A common value used is  $10k\Omega$  (see Figure 6).

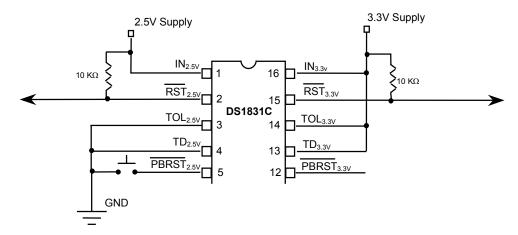
### **DS1831E APPLICATION EXAMPLE** Figure 6



### **OUTPUT VALID CONDITIONS**

The DS1831 can maintain valid outputs as long as one input remains above 1.0V. Accurate voltage monitoring additionally requires that either the 3.3V IN or 2.5V IN input be above 1.5V. If this condition is not met and at least one of the supply inputs are at or above 1.0V all outputs are maintained in the active condition. The DS1831 requires pull-up resistors on the outputs to maintain a valid output. The value of the pull up resistor is not critical in most cases but must be set low enough to pull the output to a high state. A common pull-up resistor value used is  $10k\Omega$  (see Figure 7).

### **APPLICATION DIAGRAM—OPEN DRAIN OUTPUTS** Figure 7



NOTE: If outputs are at different voltages the outputs can not be connected to form a wired AND.

#### OPERATION—NON-MASKABLE INTERRUPT

The DS1831 has two referenced comparators (DS1831D has only one referenced comparator) that can be used to monitor upstream voltages or other system specific voltages. Each comparator is referenced to the 1.25V internal band gap reference and controls an open-drain output. When a voltage being monitored decays to the voltage sense point, the DS1831 pulses the  $\overline{\text{NMI}}$  output to the active state for a minimum 10µs. The comparator detection circuitry also has built-in hysteresis of  $100\mu\text{V}$ . The supply must be below the voltage sense point for approximately 2µs before a low  $\overline{\text{NMI}}$  will be generated. In this way, power supply noise is minimized in the monitoring function, reducing false interrupts. See Figure 8 for the non-maskable timing diagram.

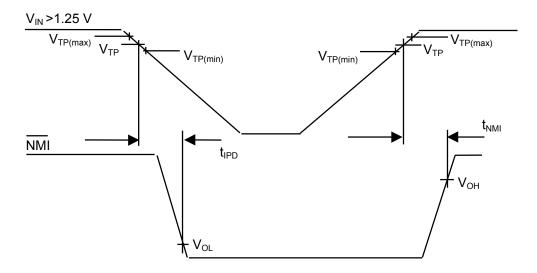
Versatile trip voltages can be configured by the use of an external resistor divider to divide the voltage at a sense point to the 1.25V trip levels of the referenced comparators. See Figure 9 for an example circuit diagram and sample equations. The equations demonstrate a design process to determine the resistor values to use.

Connecting one or both  $\overline{\text{NMI}}$  outputs to one of the reset specific  $\overline{\text{PBRST}}$  s allows the non-maskable interrupt to generate an automatic reset for the reset time period when an out-of-tolerance condition occurs in a monitored supply. An example is shown in Figure 9.

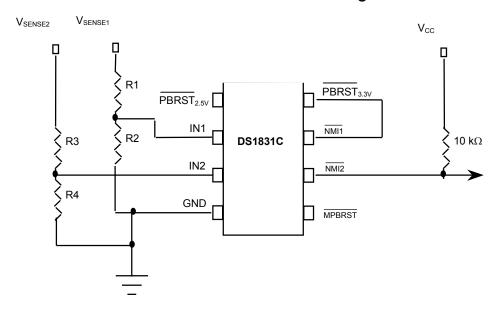
The output associated with the specific input will be held low if the voltage on the input pin is less than 1.25V. If the voltage is above 1.25V the output will not sink current and will be pulled up by the required pull-up resistor. The value of the resistors is not critical in most cases but must be set low enough to pull the output to a high state. A common value used is  $10k\Omega$ . If an  $\overline{\text{NMI}}$  output is connected to a pushbutton input an additional pull-up resistor can be used (to improve speed of transitions) but is not required.

During a power-up, any detected IN pin levels above  $V_{TP}$  by the comparator are disabled from generating an inactive (high) interrupt until at least one supply on the  $V_{IN}$  inputs rises above 1.5V. All NMI outputs will be held active (low) until at least one  $V_{IN}$  reaches 1.5V at which point the  $\overline{NMI}$  outputs will be based on the value of the associated IN input.

# TIMING DIAGRAM—NON-MASKABLE INTERRUPT Figure 8



### NON-MASKABLE INTERRUPT CIRCUIT EXAMPLE Figure 9



Example: 
$$V_{SENSE1} = 11.50 \text{ volts trip point}$$
  $V_{SENSE1} = \frac{R1 + R2}{R2}$  X 1.25V

Therefore: 
$$11.50V = \frac{R1 + 100 \text{ k}\Omega}{100 \text{ k}\Omega} \times 1.25V$$

Resulting In:  $R1 = 820 \text{ k}\Omega$ 

Repeat the same steps to solve for R3 and R4 with  $V_{\text{SENSE2}}$ .

### **OPERATION—WATCHDOG TIMER**

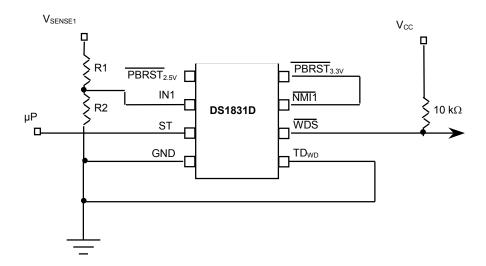
The watchdog timer function (DS1831D only) forces the  $\overline{\text{WDS}}$  signal active (low) when the  $\overline{\text{ST}}$  input does not have a transition (high-to-low or low-to-high) within the predetermined time period. The time-out period is determined by the condition of the  $TD_{WD}$  pin (see Table 1). If  $TD_{WD}$  is connected to ground the minimum watchdog time-out would be 10ms, TD floating would yield a minimum time-out of 100ms, and  $TD_{WD}$  connected to  $V_{CC}$  would provide a time-out of 1000ms minimum. Time-out of the watchdog starts when at least one of the  $\overline{\text{RST}}$  outputs becomes inactive (high). If a transition occurs on the ST input pin prior to time-out, the watchdog timer is reset and begins to time-out again. If the watchdog timer is allowed to time-out, then the  $\overline{\text{WDS}}$  output is pulsed active for a minimum of 100 $\mu$ s.

The  $\overline{\text{WDS}}$  output is an open-drain output and must be pulled up externally. In most applications this output would be connected to one of the Pushbutton inputs and would not require an external pull-up resistor. The value of the resistors is not critical in most cases but must be set low enough to pull the output to a high state. A common value used is  $10\text{k}\Omega$ . If a  $\overline{\text{WDS}}$  output is connected to a pushbutton input an additional pull-up resistor can be used (to improve speed of transitions) but is not required.

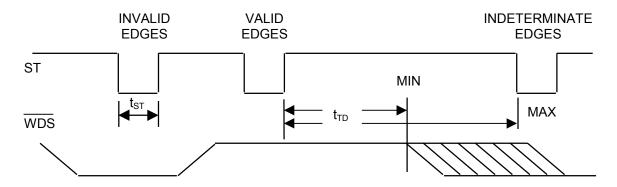
The ST input can be derived from many microprocessor outputs. The most typical signals used are the microprocessor address signals, data signals, or control signals. When the microprocessor functions normally, these signals would, as a matter of routine, cause the watchdog to be reset prior to time-out. To guarantee that the watchdog timer does not time-out, a transition must occur at or less than the minimum times shown in Table 1. A typical circuit example is shown in Figure 10. The watchdog timing is shown in Figure 11.

The DS1831A watchdog function cannot be disabled. The watchdog strobe input must be strobed to avoid a watchdog time-out however the watchdog status output can be disconnected yielding the same result.

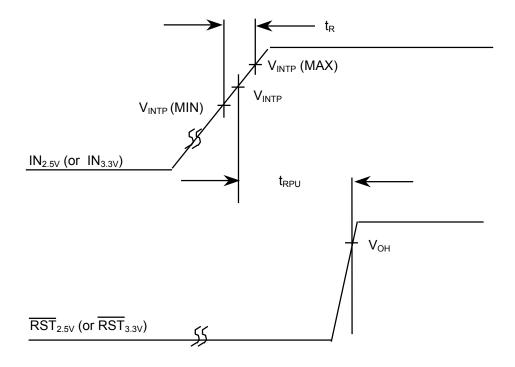
### WATCHDOG CIRCUIT EXAMPLE Figure 10



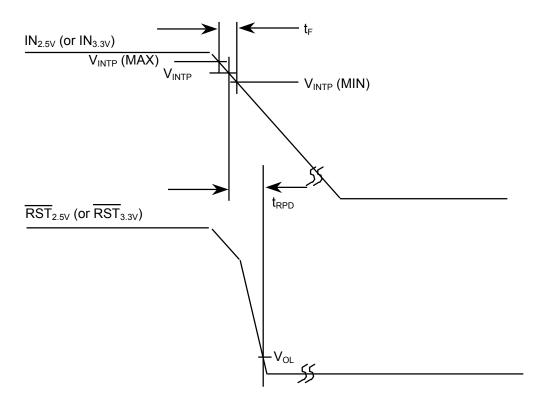
# TIMING DIAGRAM—STROBE INPUT Figure 11



# **RESET TIMING DIAGRAM—POWER UP** Figure 12



# RESET TIMING DIAGRAM—POWER DOWN Figure 13



### **ABSOLUTE MAXIMUM RATINGS\***

Voltage on  $IN_{2.5V}$  or  $IN_{3.3V}$ 

Pins Relative to Ground -0.5V to +6.0V

Voltage on either  $\overline{RST}$  Relative to Ground -0.5V to the greater of  $IN_{2.5V} + 0.5V$  or  $IN_{3.3V} + 0.5V$ 

Voltage on  $\overline{PBRST}_{3.3V}$  Relative to Ground -0.5V to  $IN_{3.3V} + 0.5V$  Voltage on  $\overline{PBRST}_{2.5V}$  Relative to Ground -0.5V to  $IN_{2.5V} + 0.5V$ 

Voltage on MPBRST, IN1, IN2

Relative to Ground -0.5V to the greater of  $IN_{2.5V} + 0.5V$  or  $IN_{3.3V} + 0.5V$ 

Operating Temperature Range -40°C to +85°C Storage Temperature Range -55°C to +125°C

Soldering Temperature See IPC/JEDEC J-STD-020A specification

#### RECOMMENDED DC OPERATING CONDITIONS

(-40°C to +85°C)

PARAMETER	SYMBOL	MIN	MAX	UNITS	NOTES
IN <sub>2.5V</sub> (Supply Voltage)	V <sub>IN</sub>	1.0	5.5	V	1
IN <sub>3.3V</sub> (Supply Voltage)	V <sub>IN</sub>	1.0	5.5	V	1
$\overline{\text{PBRST}}_{2.5\text{V}}$ , $\overline{\text{PBRST}}_{3.3\text{V}}$ , $\overline{\text{MPBRST}}$ , $\overline{\text{ST}}$ input High Level	$V_{IH}$	0.7 x V <sub>INT</sub>	$V_{INT} + 0.3$	V	1*
PBRST 3.3V, PBRST 5V, MPBRST, ST input Low Level	$V_{IL}$	-0.3	$0.3 \times V_{INT}$	V	1*

<sup>\*</sup>  $V_{INT}$  is the greater voltage level of the  $IN_{2.5V}$  or  $IN_{3.3V}$ .

#### DC ELECTRICAL CHARACTERISTICS

 $(-40^{\circ}\text{C to }85^{\circ}\text{C}; IN_{2.5\text{V}}, IN_{3.3\text{V}} = 1.0\text{V to }5.5\text{V})$ 

PARAMETER	SYMBOL	MIN	TYP	MAX	UNITS	NOTES
Input Leakage	$I_{\mathrm{IL}}$	-1.0		+1.0	μΑ	2
Output Current @ 2.4V	$I_{OH}$					3
Output Current @ 0.4V	$I_{OL}$	+10			mA	4
Operating Current @ ←5.5V	$I_{CC}$		80	100	μΑ	5
Operating Current @ ←3.6V	$I_{CC}$		60	85	μΑ	6
$IN_{3.3V}$ Trip Point (TOL <sub>3.3V</sub> = $IN_{3.3V}$ )	V <sub>INTP</sub>	2.98	3.06	3.15	V	
$IN_{3.3V}$ Trip Point (TOL <sub>3.3V</sub> = GND)	V <sub>INTP</sub>	2.80	2.88	2.97	V	
$IN_{3.3V}$ Trip Point (TOL <sub>3.3V</sub> = Float)	$V_{INTP}$	2.47	2.55	2.64	V	
$IN_{2.5V}$ Trip Point ( $TOL_{2.5V} = IN_{2.5V}$ )	V <sub>INTP</sub>	2.250	2.312	2.375	V	
$IN_{2.5V}$ Trip Point (TOL <sub>2.5V</sub> = GND)	$V_{INTP}$	2.125	2.187	2.250	V	
$IN_{2.5V}$ Trip Point (TOL <sub>2.5V</sub> = Float)	V <sub>INTP</sub>	2.000	2.062	2.125	V	
IN Input Trip Points	$V_{TP}$	1.15	1.25	1.30	V	

<sup>\*</sup> This is a stress rating only and functional operation of the device at these or any other conditions above those indicated in the operation sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods of time may affect reliability.

### CAPACITANCE

 $(t_A = +25^{\circ}C)$ 

PARAMETER	SYMBOL	MAX	UNITS	NOTES
Input Capacitance	$C_{IN}$	5	pF	
Input Capacitance	$C_{OUT}$	7	pF	

### **AC ELECTRICAL CHARACTERISTICS**

 $(-40^{\circ} \text{ to } 85^{\circ}\text{C}; \text{ IN}_{2.5\text{V}}, \text{ IN}_{3.3\text{V}} = 1.0\text{V to } 5.5\text{V})$ 

PARAMETER	SYMBOL	MIN	TYP	MAX	UNITS	NOTES
RESET Active Time (TD=Low)	$t_{RST}$	10	16	20	ms	6
RESET Active Time (TD=Float)	$t_{RST}$	100	160	200	ms	6
RESET Active Time (TD=High)	$t_{RST}$	1000	1600	2000	ms	6
V <sub>CC</sub> Detect to RST	$t_{ m RPU}$	See R	ESET Activ	e Time	ms	6
V <sub>CC</sub> Detect to RST	$t_{RPD}$		2	10	μs	7
VIN Detect to NMI	$t_{\mathrm{IPD}}$		2	10	μs	7
NMI Active Time	t <sub>NMI</sub>	20			μs	
$\overline{PBRST} = V_{IL}$	$t_{\mathrm{PB}}$	20			μs	
PBRST Stable Low to Reset Active	$t_{ m PDLY}$			50	μs	
Watchdog Timeout (TD <sub>(WD)</sub> =Low)	$t_{TD}$	10	16	20	ms	
Watchdog Timeout (TD <sub>WD</sub> =Float)	$t_{TD}$	100	160	200	ms	
Watchdog Timeout (TD <sub>WD</sub> =High)	$t_{TD}$	1000	1600	2000	ms	
ST Pulse Width	$t_{\mathrm{ST}}$	10			ns	
$V_{in}$ Slew Rate (V $_{INTP(MAX)}$ to V $_{INTP(MIN)}$ )	$t_{ m F}$	300			μs	
$V_{in}$ Slew Rate (V $_{INTP(MAX)}$ to V $_{INTP(MIN)}$ )	$t_{\mathrm{R}}$	0			ns	

### **NOTES:**

- 1) All voltages are referenced to ground.
- 2) All Pushbutton inputs are internally pulled to the associated Supply IN input or the greatest Supply IN input for the  $\overline{\text{MPBRST}}$  with an internal Impedance of  $100\text{k}\Omega$ .
- 3) All outputs are Open Drain and output I<sub>OH</sub> would be determined by the external pull-up resistor.
- 4) Measured with outputs open and  $IN_{3.3V}$  or  $IN_{2.5V} \le 5.5V$ .
- 5) Measured with outputs open and  $IN_{3.3V}$  or  $IN_{2.5V} \le 3.6V$ .
- 6) Measured using  $t_R = 5 \mu s$ .
- 7) Noise immunity pulses  $<2\mu s$  at a trip level will not cause a  $\overline{RST}$  or  $\overline{NMI}$ .

# **ORDERING INFORMATION**

<b>Ordering Part</b>	Package Type	Description	
Number			
DS1831C	16-Pin DIP 300mil	2.5V/3.3V Multisupply Monitor	
DS1831CS	16-Pin SO 150mil	2.5V/3.3V Multisupply Monitor	
DS1831D	16-Pin DIP 300mil	2.5V/3.3V Multisupply Monitor w/Watchdog	
DS1831DS	16-Pin SO 150mil	2.5V/3.3V Multisupply Monitor w/Watchdog	
DS1831E	16-Pin DIP 300mil	2.5V/3.3V Multisupply Monitor w/Pushbutton Status	
DS1831ES	16-Pin SO 150mil	2.5V/3.3V Multisupply Monitor w/Pushbutton Status	

<sup>\*</sup> Add "/T&R" for tape and reeling of surface mount packages.